

21.17: Power Handling Capability of Coaxial Cable

The term “power handling” refers to maximum power that can be safely transferred by a transmission line. This power is limited because when the electric field becomes too large, dielectric breakdown and arcing may occur. This may result in damage to the line and connected devices, and so must be avoided. Let E_{pk} be the maximum safe value of the electric field intensity within the line, and let P_{max} be the power that is being transferred under this condition. This section addresses the following question: How does one design a coaxial cable to maximize P_{max} for a given E_{pk} ?

We begin by finding the electric potential V within the cable. This can be done using Laplace’s equation:

$$\nabla^2 V = 0$$

Using the cylindrical (ρ, ϕ, z) coordinate system with the z axis along the inner conductor, we have $\partial V / \partial \phi = 0$ due to symmetry. Also we set $\partial V / \partial z = 0$ since the result should not depend on z . Thus, we have:

$$\frac{1}{\rho} \frac{\partial}{\partial \rho} \left(\rho \frac{\partial V}{\partial \rho} \right) = 0$$

Solving for V , we have

$$V(\rho) = A \ln \rho + B$$

where A and B are arbitrary constants, presumably determined by boundary conditions. Let us assume a voltage V_0 measured from the inner conductor (serving as the “+” terminal) to the outer conductor (serving as the “−” terminal). For this choice, we have:

$$V(a) = V_0 \rightarrow A \ln a + B = V_0 \quad (21.17.1)$$

$$V(b) = 0 \rightarrow A \ln b + B = 0 \quad (21.17.2)$$

Subtracting the second equation from the first and solving for A , we find $A = -V_0 / \ln(b/a)$. Subsequently, B is found to be $V_0 \ln(b) / \ln(b/a)$, and so

$$V(\rho) = \frac{-V_0}{\ln(b/a)} \ln \rho + \frac{V_0 \ln(b)}{\ln(b/a)}$$

The electric field intensity is given by:

$$\mathbf{E} = -\nabla V$$

Again we have $\partial V / \partial \phi = \partial V / \partial z = 0$, so

$$\mathbf{E} = -\hat{\rho} \frac{\partial}{\partial \rho} V \quad (21.17.3)$$

$$= -\hat{\rho} \frac{\partial}{\partial \rho} \left[\frac{-V_0}{\ln(b/a)} \ln \rho + \frac{V_0 \ln(b)}{\ln(b/a)} \right] \quad (21.17.4)$$

$$= +\hat{\rho} \frac{V_0}{\rho \ln(b/a)} \quad (21.17.5)$$

Note that the maximum electric field intensity in the spacer occurs at $\rho = a$; i.e., at the surface of the inner conductor. Therefore:

$$E_{pk} = \frac{V_0}{a \ln(b/a)}$$

The power transferred by the line is maximized when the impedances of the source and load are matched to Z_0 . In this case, the power transferred is $V_0^2 / 2Z_0$. Recall that the characteristic impedance Z_0 is given in the “low-loss” case as

$$Z_0 \approx \frac{1}{2\pi} \frac{\eta_0}{\sqrt{\epsilon_r}} \ln \left(\frac{b}{a} \right) \quad (21.17.6)$$

Therefore, the maximum safe power is

$$P_{max} = \frac{V_0^2}{2Z_0} \quad (21.17.7)$$

$$\approx \frac{E_{pk}^2 a^2 \ln^2(b/a)}{2 \cdot (1/2\pi) (\eta_0/\sqrt{\epsilon_r}) \ln(b/a)} \quad (21.17.8)$$

$$= \frac{\pi E_{pk}^2}{\eta_0/\sqrt{\epsilon_r}} a^2 \ln(b/a) \quad (21.17.9)$$

Now let us consider if there is a value of a which maximizes P_{max} . We do this by seeing if $\partial P_{max}/\partial a = 0$ for some values of a and b . The derivative is worked out in an addendum at the end of this section. Using the result from the addendum, we find:

$$\frac{\partial}{\partial a} P_{max} = \frac{\pi E_{pk}^2}{\eta_0/\sqrt{\epsilon_r}} [2a \ln(b/a) - a] \quad (21.17.10)$$

For the above expression to be zero, it must be true that $2 \ln(b/a) - 1 = 0$. Solving for b/a , we obtain:

$$\frac{b}{a} = \sqrt{e} \cong 1.65 \quad (21.17.11)$$

for optimum power handling. In other words, 1.65 is the ratio of the radii of the outer and inner conductors that maximizes the power that can be safely handled by the cable.

Equation 21.17.9 suggests that ϵ_r should be maximized in order to maximize power handling, and you wouldn't be wrong for noting that, however, there are some other factors that may indicate otherwise. For example, a material with higher ϵ_r may also have higher σ_s , which means more current flowing through the spacer and thus more ohmic heating. This problem is so severe that cables that handle high RF power often use air as the spacer, even though it has the *lowest* possible value of ϵ_r . Also worth noting is that σ_{ic} and σ_{oc} do not matter according to the analysis we've just done; however, to the extent that limited conductivity results in significant ohmic heating in the conductors – which we have also not considered – there may be something to consider. Suffice it to say, the actionable finding here concerns the ratio of the radii; the other parameters have not been suitably constrained by this analysis.

Substituting \sqrt{e} for b/a in Equation 21.17.6 we find:

$$Z_0 \approx \frac{30.0 \Omega}{\sqrt{\epsilon_r}}$$

This is the characteristic impedance of coaxial line that optimizes power handling, subject to the caveats identified above. For air-filled cables, we obtain 30 Ω . Since $\epsilon_r \geq 1$, this optimum impedance is less for dielectric-filled cables than it is for air-filled cables.

Summarizing:

The power handling capability of coaxial transmission line is optimized when the ratio of radii of the outer to inner conductors b/a is about 1.65. For the air-filled cables typically used in high-power applications, this corresponds to a characteristic impedance of about 30 Ω .

Addendum: Derivative of $(1/a + C/b) / \ln(b/a)$

Evaluation of Equation 21.17.9 requires finding the derivative of $(1/a + C/b) / \ln(b/a)$ with respect to a . Using the chain rule, we find:

$$\begin{aligned} \frac{\partial}{\partial a} \left[\frac{1/a + C/b}{\ln(b/a)} \right] &= \left[\frac{\partial}{\partial a} \left(\frac{1}{a} + \frac{C}{b} \right) \right] \ln^{-1} \left(\frac{b}{a} \right) \\ &\quad + \left(\frac{1}{a} + \frac{C}{b} \right) \left[\frac{\partial}{\partial a} \ln^{-1} \left(\frac{b}{a} \right) \right] \end{aligned} \quad (21.17.12)$$

Note

$$\frac{\partial}{\partial a} \left(\frac{1}{a} + \frac{C}{b} \right) = -\frac{1}{a^2}$$

To handle the quantity in the second set of square brackets, first define $v = \ln u$, where $u = b/a$. Then:

$$\begin{aligned} \frac{\partial}{\partial a} v^{-1} &= \left[\frac{\partial}{\partial v} v^{-1} \right] \left[\frac{\partial v}{\partial u} \right] \left[\frac{\partial u}{\partial a} \right] \\ &= [-v^{-2}] \left[\frac{1}{u} \right] [-ba^{-2}] \\ &= \left[-\ln^{-2} \left(\frac{b}{a} \right) \right] \left[\frac{a}{b} \right] [-ba^{-2}] \\ &= \frac{1}{a} \ln^{-2} \left(\frac{b}{a} \right) \end{aligned} \quad (21.17.13)$$

So:

$$\begin{aligned} \frac{\partial}{\partial a} \left[\frac{1/a + C/b}{\ln(b/a)} \right] &= \left[-\frac{1}{a^2} \right] \ln^{-1} \left(\frac{b}{a} \right) \\ &\quad + \left(\frac{1}{a} + \frac{C}{b} \right) \left[\frac{1}{a} \ln^{-2} \left(\frac{b}{a} \right) \right] \end{aligned} \quad (21.17.14)$$

This result is substituted in Equation 21.17.9 to obtain Equation 21.17.10

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